

TEN 20-WIN Series

Application Note

DC/DC Converter 9 to 36Vdc or 18 to 75 Vdc Input 3.3 to 15Vdc Single Outputs and ±5 to ±15Vdc Dual Outputs, 20W



Complete TEN 20-WIN datasheet can be downloaded at: http://www.tracopower.com/products/ten20WIN.pdf

Features

- Single output up to 5.5A
 Dual output up to ±2.0A
- 20 watts maximum output power
- 4:1 ultra wide input voltage range of 9-36 and 18-75VDC
- Six-sided continuous shield
- Case grounding
- High efficiency up to 89%
- Low profile: 50.8×25.4×10.2 mm (2.00×1.00×0.40 inch)
- Fixed switching frequency
- RoHS directive compliant
- No minimum load
- Input to output isolation: 1500Vdc for 1 minute
- Operating case temperature range: 105°C max
- Input under-voltage protection
- Output over-voltage protection
- Over-current protection, auto-recovery
- Output short circuit protection
- Remote on/off

Options

Heat sinks available for extended operation

Applications

- Distributed power architectures
- Computer equipment
- Communications equipment

General Description

The TEN 20-WIN series offer 20 watts of output power from a 50.8×25.4×10.2mm package with a 4:1 ultra wide input voltage of 9~36Vdc, 18~75Vdc. The product features 1500VDC of isolation, short circuit and over voltage protection, as well as six sided shielding. All models are particularly suited to telecommunications, industrial, mobile telecom and test equipment applications.

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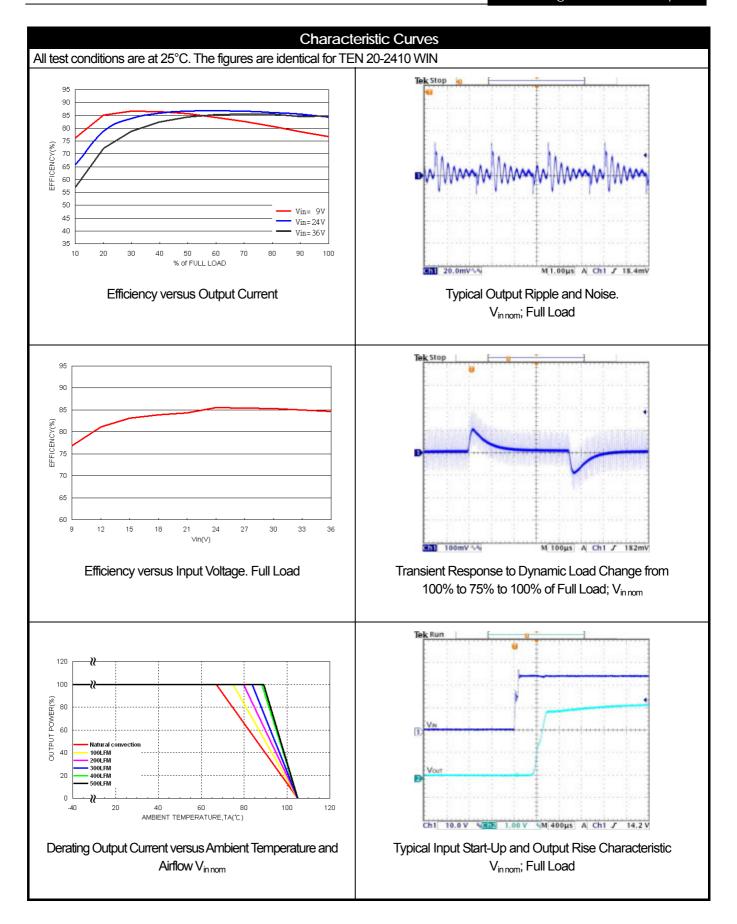
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Absolute Maximum Rating						
Parameter Model Min Max Ui						
Input Voltage						
Continuous	TEN 20-24xx WIN		40			
	TEN 20-48xx WIN		80	Vdc		
Transient (100mS)	TEN 20-24xx WIN		50			
, , ,	TEN 20-48xx WIN		100			
Input Voltage Variation (complies with ETS300 132 part 4.4)	All		5	V/mS		
Operating Ambient Temperature (with derating)	All	-40	105	°C		
Operating Case Temperature	All		105	°C		
Storage Temperature	All	-55	125	°C		

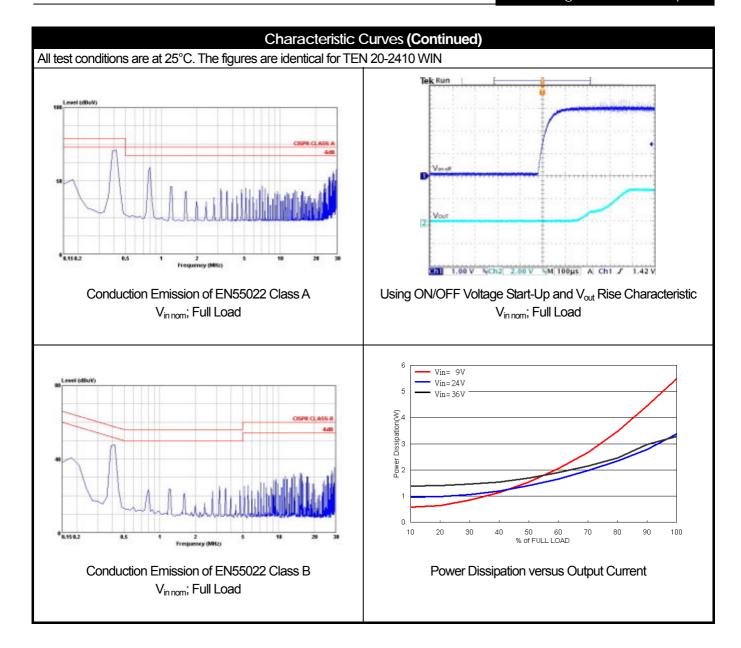
O	utput Specification				
Parameter	Model	Min	Тур	Max	Unit
Output Voltage Range	TEN 20-xx10 WIN	3.267	3.3	3.333	
$(V_{in nom}; Full Load; T_A = 25^{\circ}C)$	TEN 20-xx11 WIN	4.95	5.0	5.05	
	TEN 20-xx12 WIN	11.88	12.0	12.12	
	TEN 20-xx13 WIN	14.85	15.0	15.15	Vdc
	TEN 20-xx21 WIN	±4.95	±5.0	±5.05	
	TEN 20-xx22 WIN	±11.88	±12.0	±12.12	
	TEN 20-xx23 WIN	±14.85	±15.0	±15.15	
Output Regulation					
Line (V _{in min} to V _{in max} at Full Load)	All	-0.2		+0.2	%
Load (0% to 100% of Full Load)		-0.5		+0.5	
Output Ripple & Noise	TEN 20-xx10			60	
Peak-to-Peak (20MHz bandwidth)	Others single output			75	mV _{Pk-Pk}
(Measured with a 0.1µF/50V MLCC)	All dual output			100	
Temperature Coefficient	All	-0.02		+0.02	%/°C
Output Voltage Overshoot	All		0	3	0/ \/
$(V_{in min} \text{ to } V_{in max}; \text{ Full Load}; T_A = 25^{\circ}\text{C})$	All		0	3	% Vou⊤
Dynamic Load Response					
$(V_{in nom}; T_A = 25^{\circ}C)$					
Load step change from					
75% to 100% or 100 to 75% of Full Load	All		200		mV
Peak Deviation	All		250		μS
Setting Time (V _{OUT} < 10% peak deviation)			250		μΟ
Output Current	TEN 20-xx10 WIN	0		5500	
	TEN 20-xx11 WIN	0		4000	
	TEN 20-xx12 WIN	0		1670	
	TEN 20-xx13 WIN	0		1330	mA
	TEN 20-xx21 WIN	0		±2000	
	TEN 20-xx22 WIN	0		±833	
	TEN 20-xx23 WIN	0		±667	
Output Over Voltage Protection	TEN 20-xx10 WIN		3.9		
(Zener diode clamp)	TEN 20-xx11 WIN		6.2		
	TEN 20-xx12 WIN		15		
	TEN 20-xx13 WIN		18		Vdc
	TEN 20-xx21 WIN		6.2		
	TEN 20-xx22 WIN		15		
	TEN 20-xx23 WIN		18		
Output Over Current Protection	All		150		% FL.
Output Short Circuit Protection	All	Hic	ccup, automa	atics recover	У

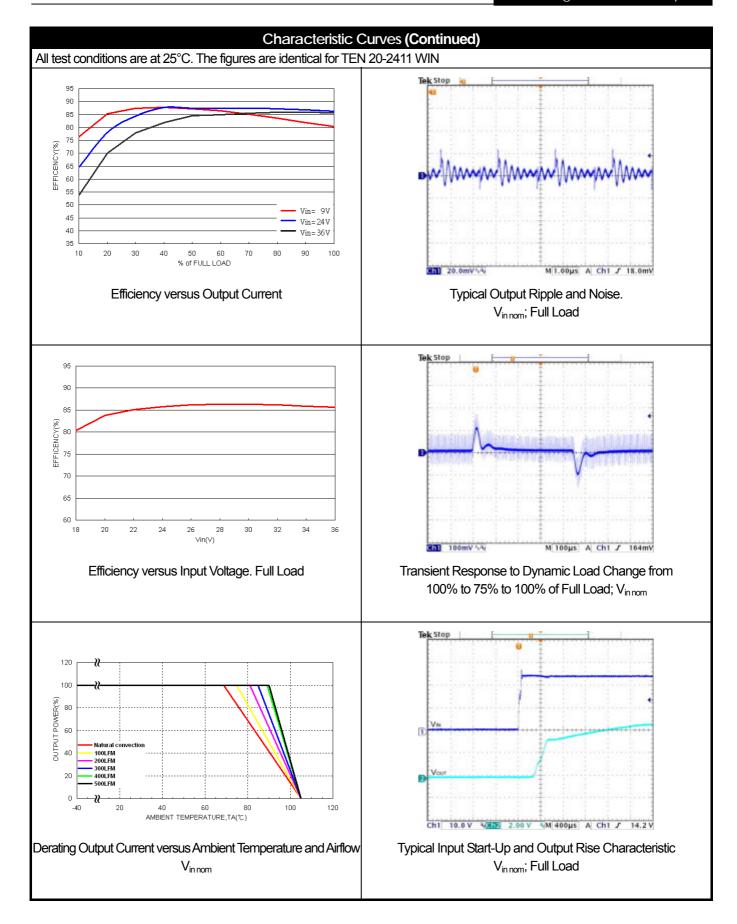
	Input Specification				
Parameter	Model	Min	Тур	Max	Unit
Operating Input Voltage	TEN 20-24xx WIN	9	24	36	Vdc
	TEN 20-48xx WIN	18	48	75	Vuc
Input Current	TEN 20-2410 WIN			934	
(Maximum value at V _{in nom} ; Full Load)	TEN 20-2411 WIN			992	
,	TEN 20-2412 WIN			1018	
	TEN 20-2413 WIN			1014	
	TEN 20-2421 WIN			992	
	TEN 20-2422 WIN			1004	
	TEN 20-2423 WIN			1005	
	TEN 20-4810 WIN			467	mA
	TEN 20-4811 WIN			496	
	TEN 20-4812 WIN			503	
	TEN 20-4813 WIN			501	
	TEN 20-4821 WIN			490	
	TEN 20-4822 WIN			496	
	TEN 20-4823 WIN			496	
Innut Standby autont	TEN 20-2410 WIN		50	490	
Input Standby current			50		
(Typical value at V _{in nom} ; No Load)	TEN 20-2411 WIN		65		
	TEN 20-2412 WIN		22		
	TEN 20-2413 WIN		22		
	TEN 20-2421 WIN		55		
	TEN 20-2422 WIN		30		
	TEN 20-2423 WIN		30		mA
	TEN 20-4810 WIN		35		
	TEN 20-4811 WIN		35		
	TEN 20-4812 WIN		15		
	TEN 20-4813 WIN		15		
	TEN 20-4821 WIN		35		
	TEN 20-4822 WIN		17		
	TEN 20-4823 WIN		17		
Under Voltage Lockout Turn-on Threshold	TEN 20-24xx WIN		9		\/da
	TEN 20-48xx WIN		18		Vdc
Under Voltage Lockout Turn-off Threshold	TEN 20-24xx WIN		7.5		
	TEN 20-48xx WIN		15		Vdc
Input reflected ripple current					_
(5 to 20MHz, 12µH source impedance)	All		20		mA _{Pk-Pk}
Start Up Time					
(V _{in nom} and constant resistive load)					
Power up	All		20		mS
Remote On/Off	7 41		20		1110
Remote On/Off Control			20		
(The On/Off pin voltage is referenced to -V _{IN})	All	2		10	\ /al -
On/Off pin High Voltage (Remote On)		3		12	Vdc
On/Off pin Low Voltage (Remote Off)		0		1.2	
Remote Off input current	All			2.5	mA
Input current of Remote control pin	All	-0.5		0.5	mA

General Specification					
Parameter	Model	Min	Тур	Max	Unit
Efficiency	TEN 20-2410 WIN		85		
$(V_{in nom}; Full Load; T_A = 25^{\circ}C)$	TEN 20-2411 WIN		88		
	TEN 20-2412 WIN		86		
	TEN 20-2413 WIN		86		
	TEN 20-2421 WIN		88		
	TEN 20-2422 WIN		87		
	TEN 20-2423 WIN		87		%
	TEN 20-4810 WIN		85		70
	TEN 20-4811 WIN		88		
	TEN 20-4812 WIN		87		
	TEN 20-4813 WIN		87		
	TEN 20-4821 WIN		89		
	TEN 20-4822 WIN		88		
	TEN 20-4823 WIN		88		
Isolation voltage					
Input to Output (for 60 seconds)	All	1500			Vdc
Input to Case, Output to Case (for 60 seconds)		1500			
Isolation resistance	All	1			GΩ
Isolation capacitance	All			1500	рF
Switching Frequency	All		400		KHz
Weight	All		27.0		g
MTBF					
Bellcore TR-NWT-000332, T _A = +40°C	All		1'691'000		hours
MIL-STD-217F, $T_A = +25^{\circ}C$			562'900		

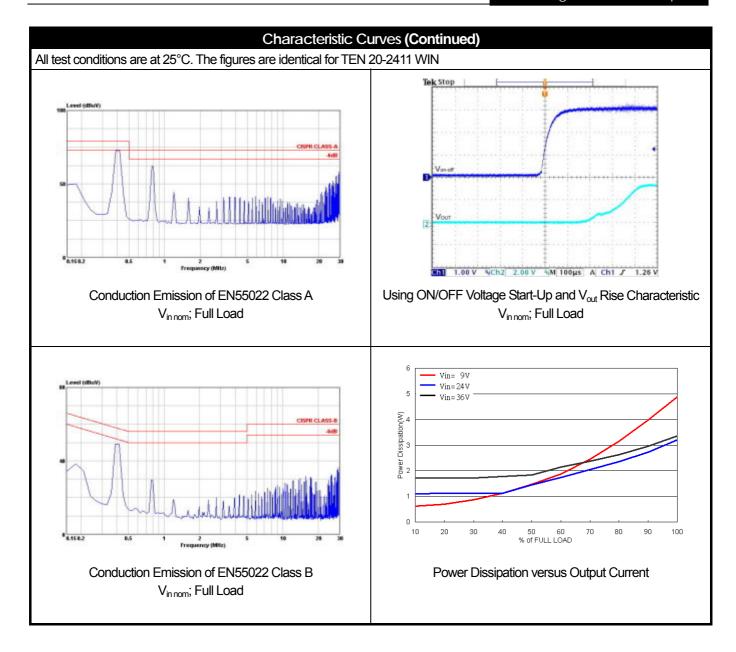


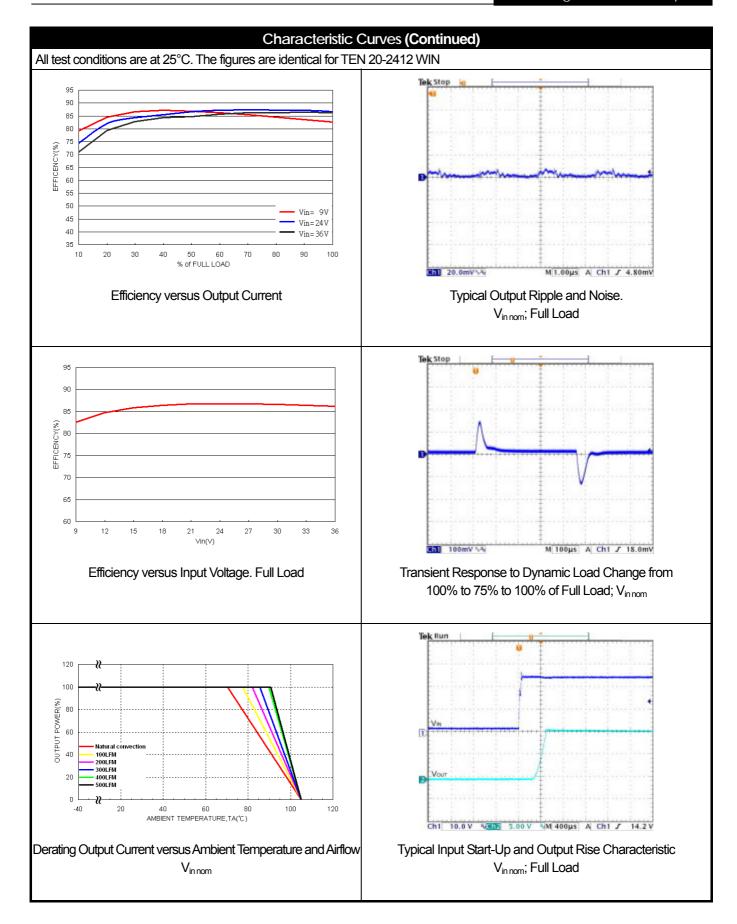
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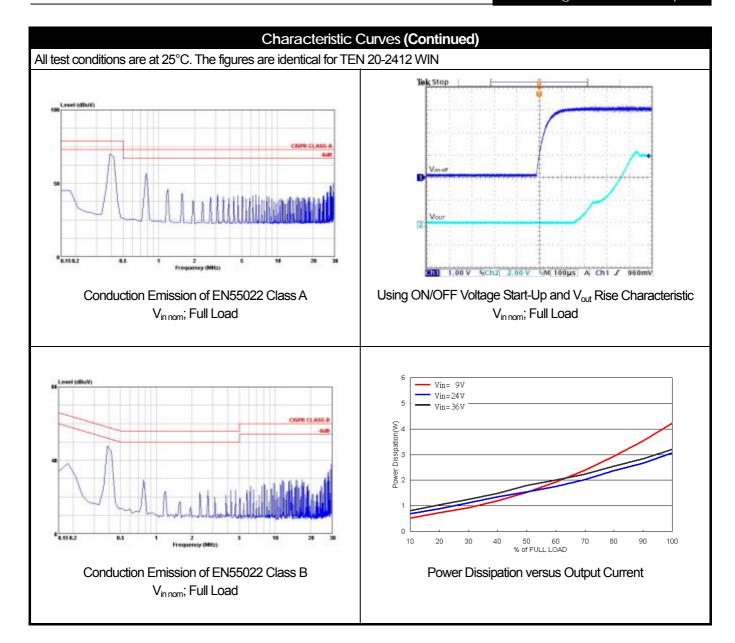


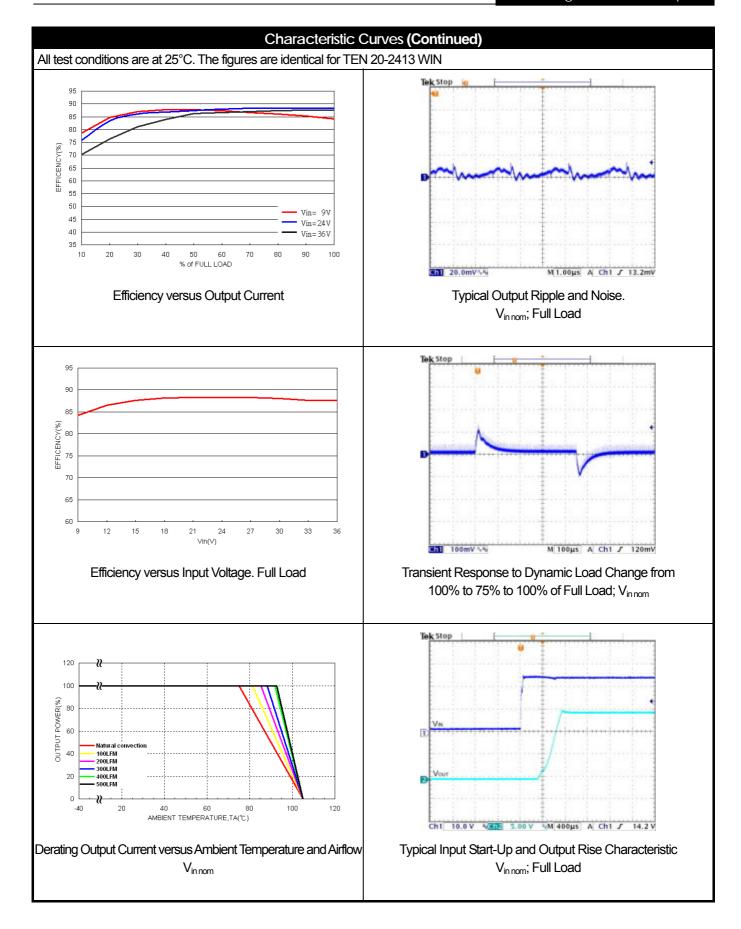


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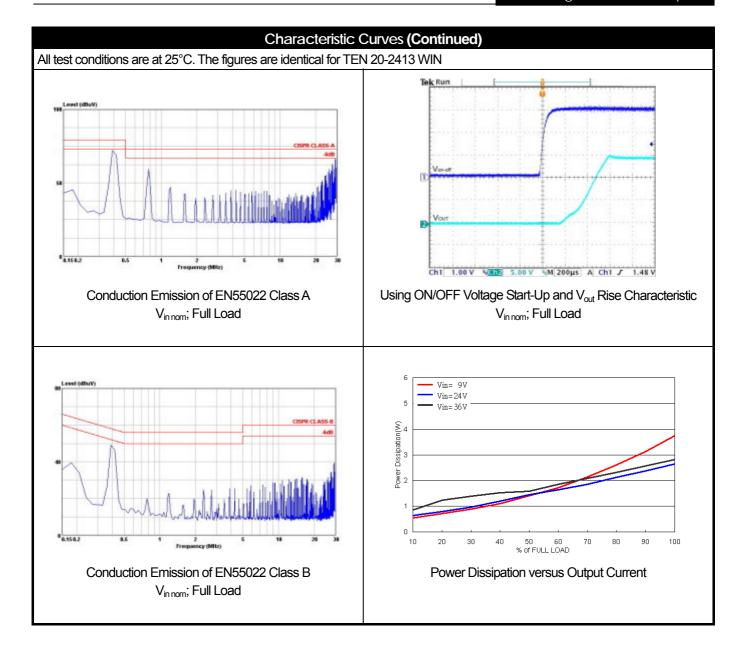


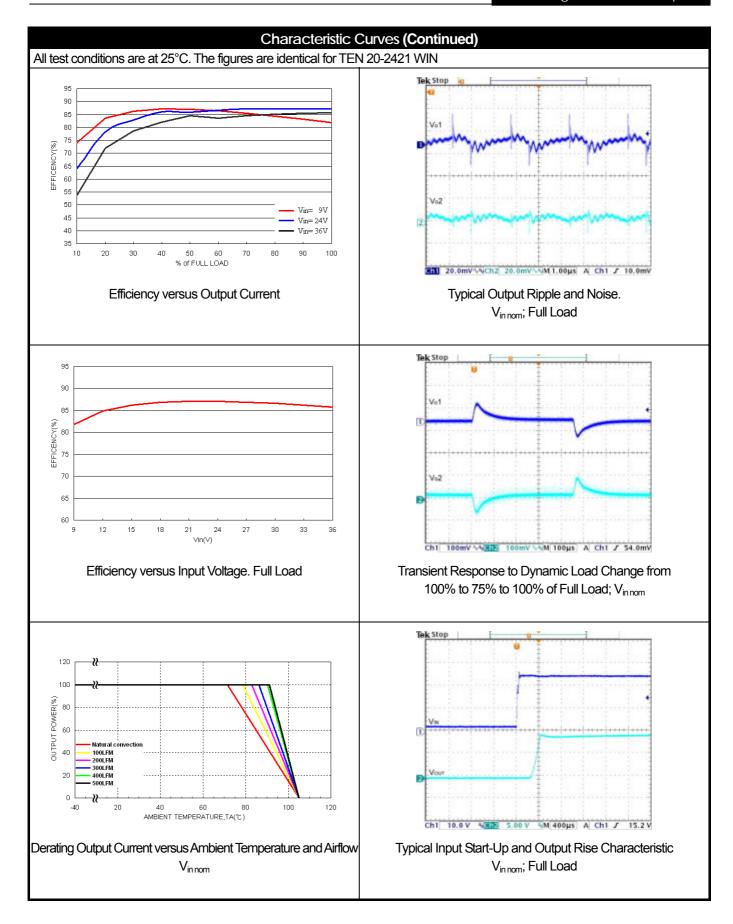




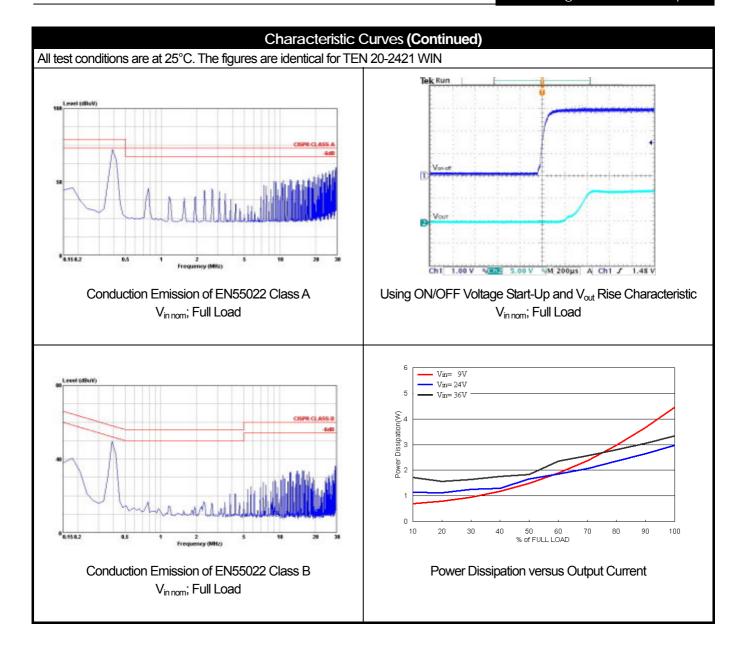


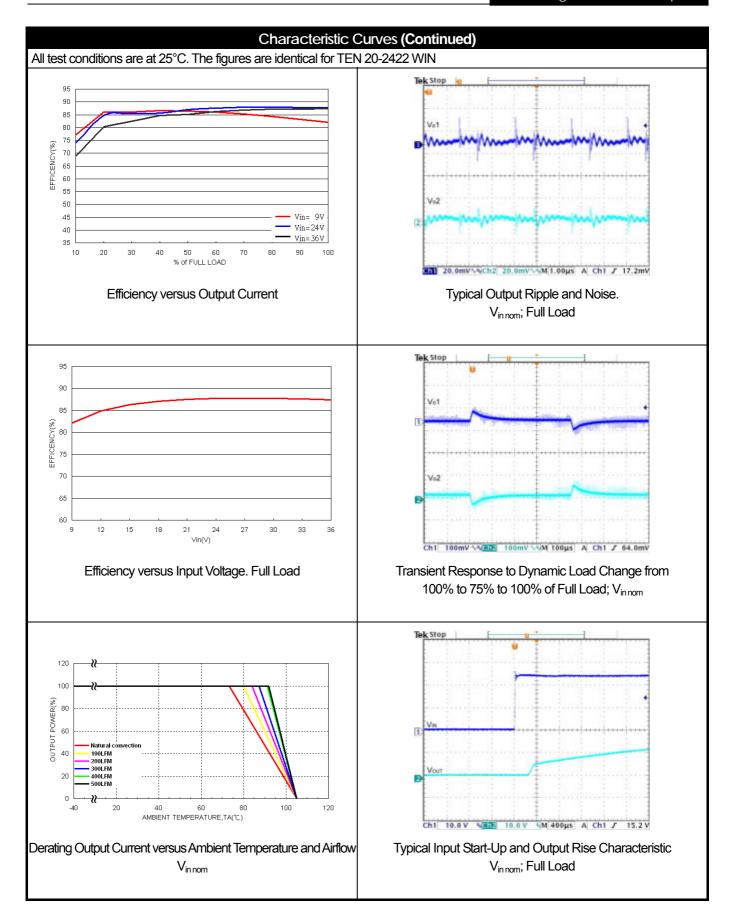
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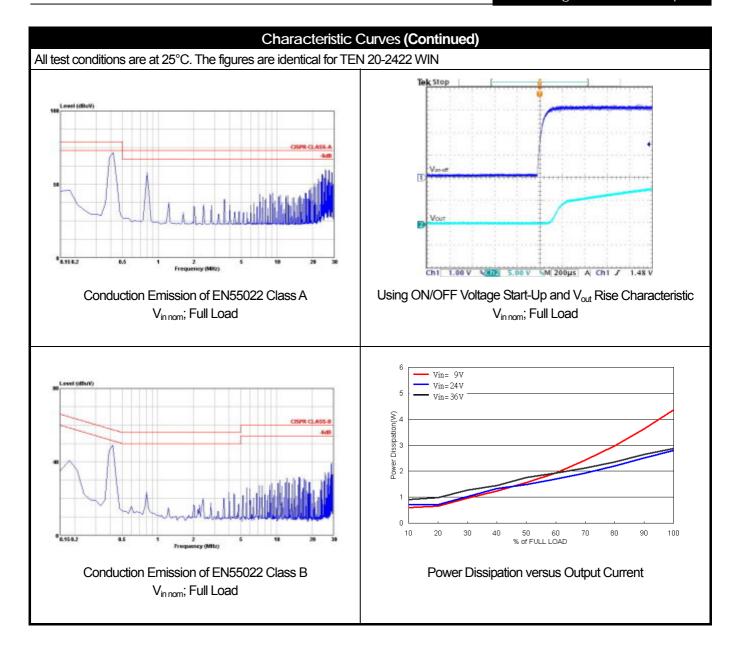


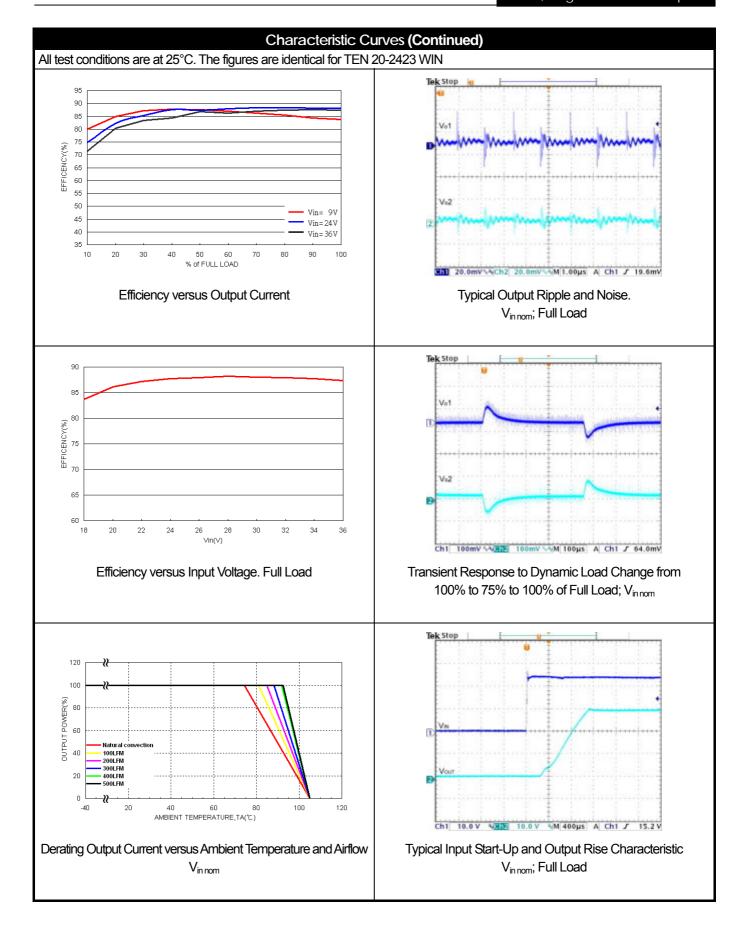
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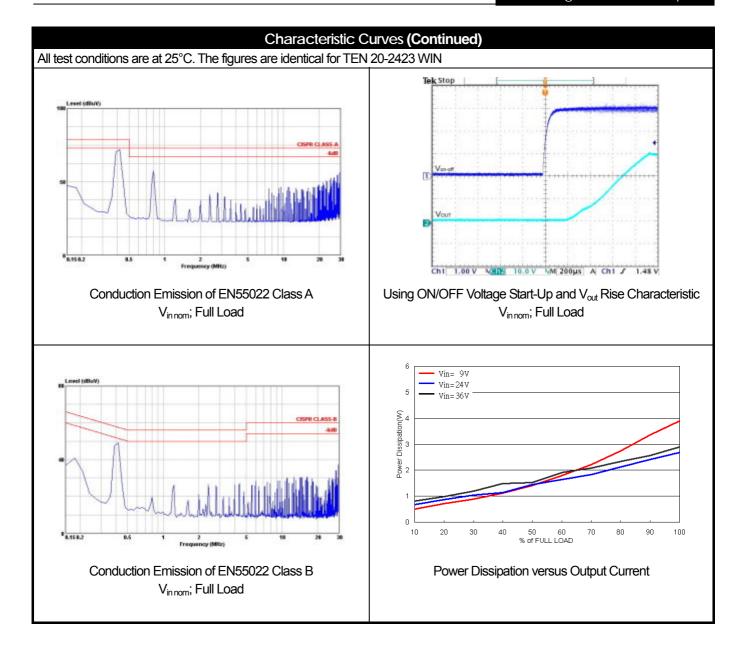


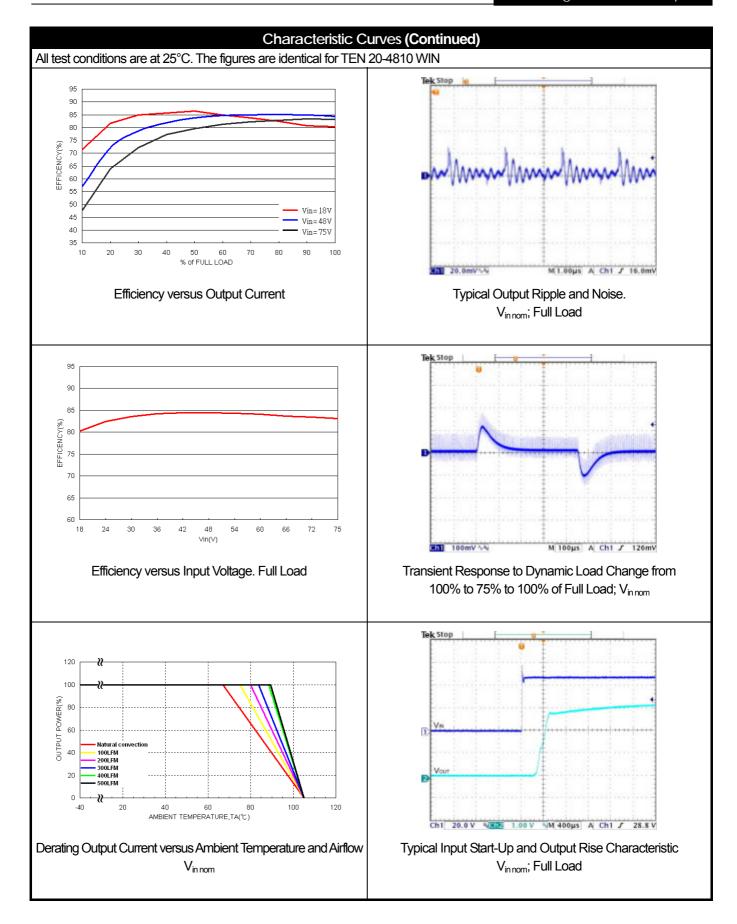
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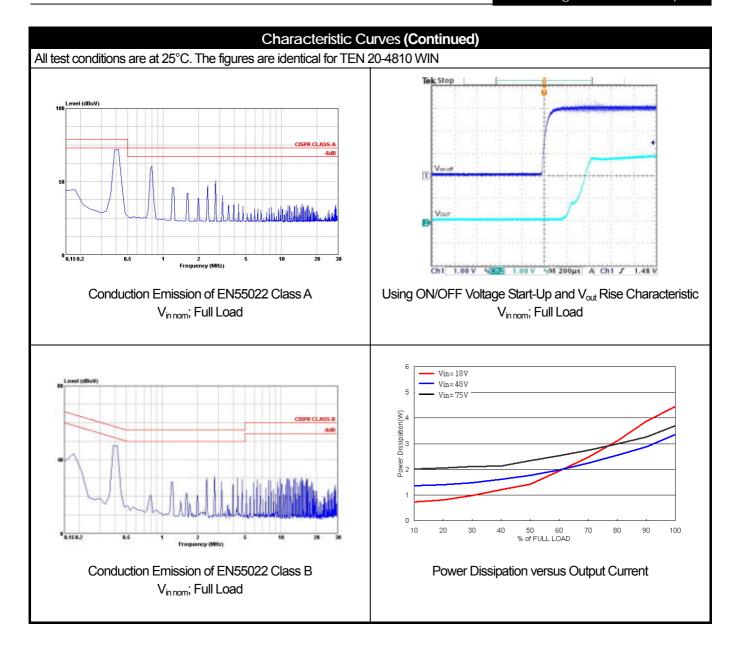


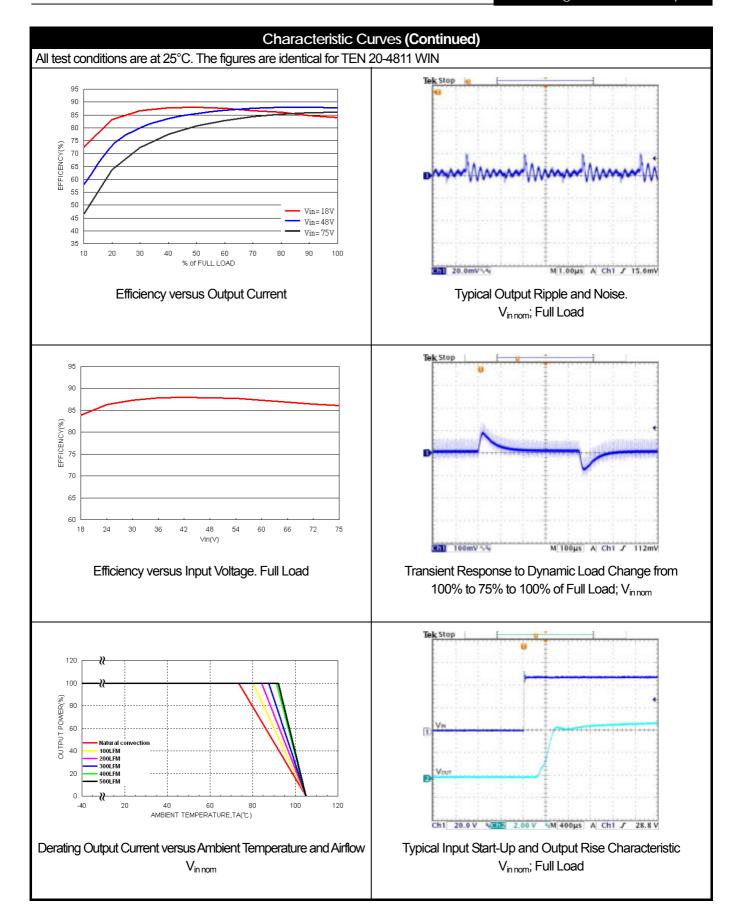
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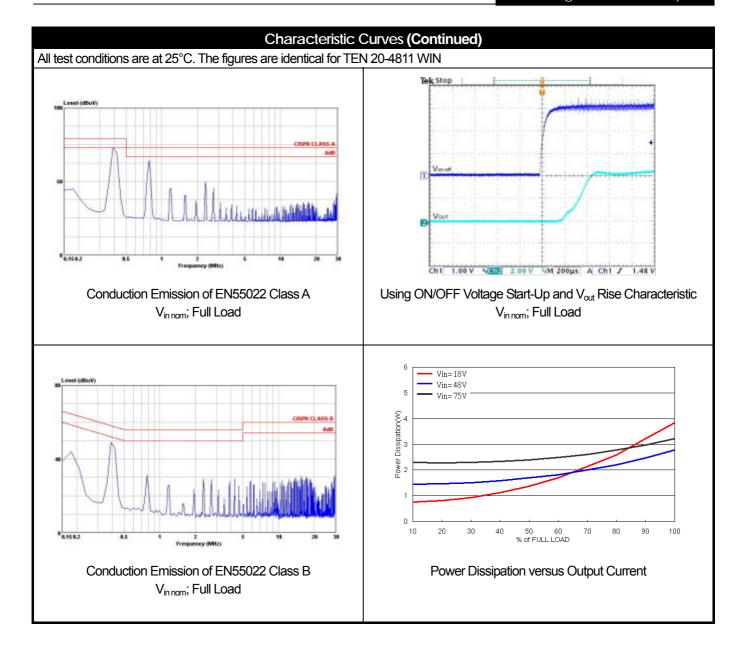


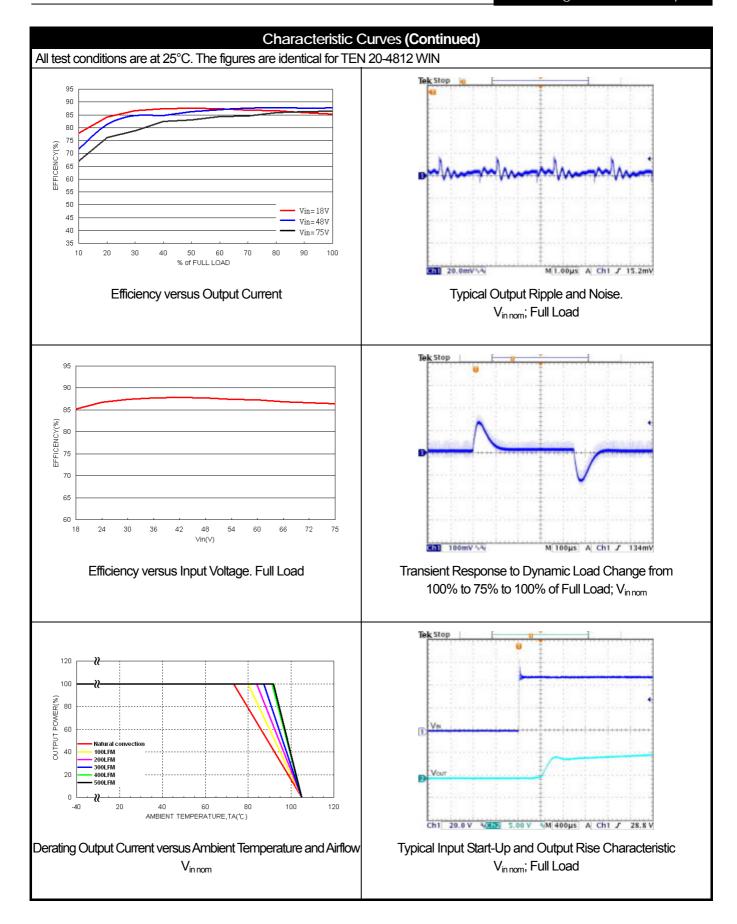
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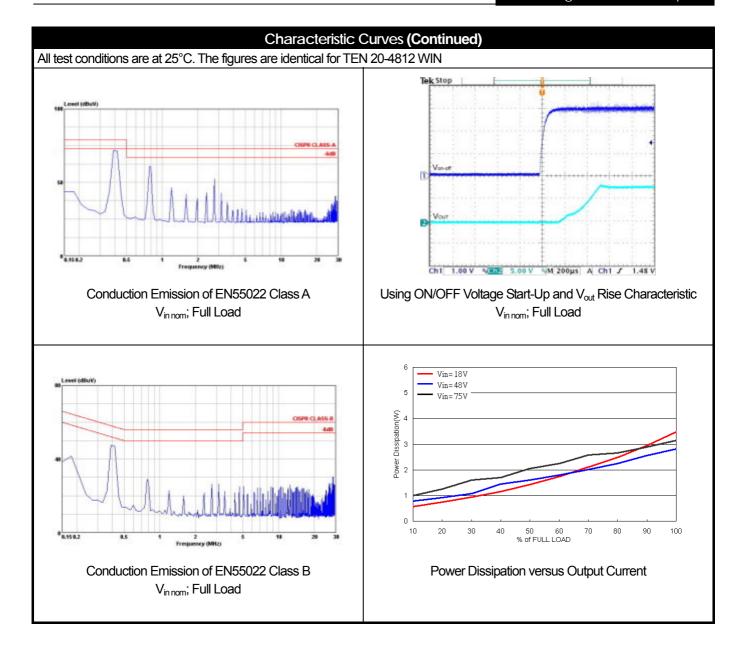


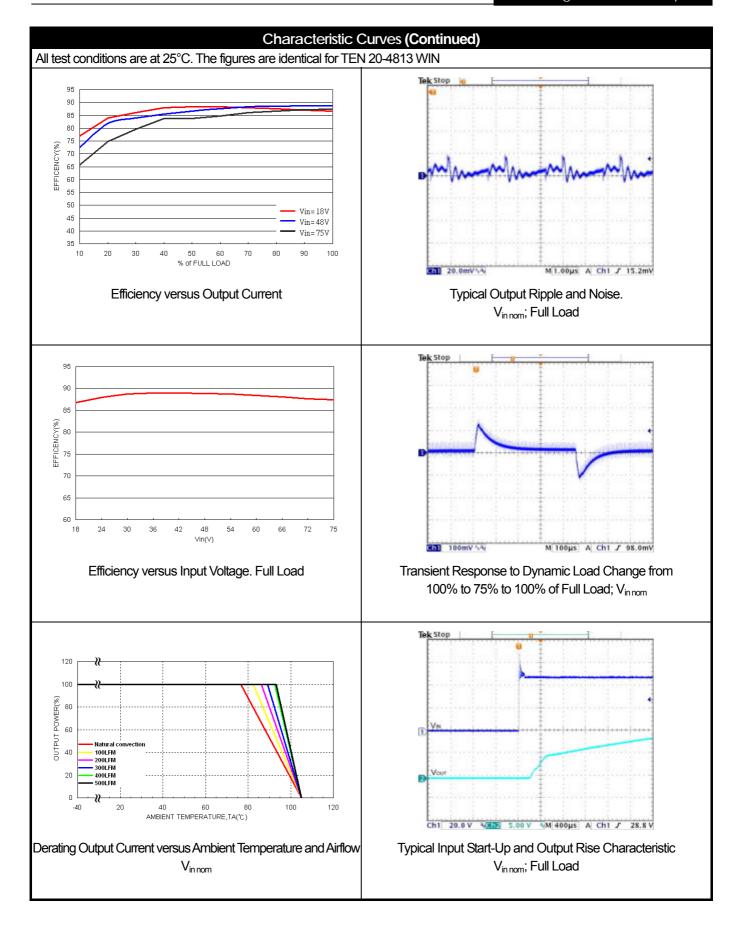
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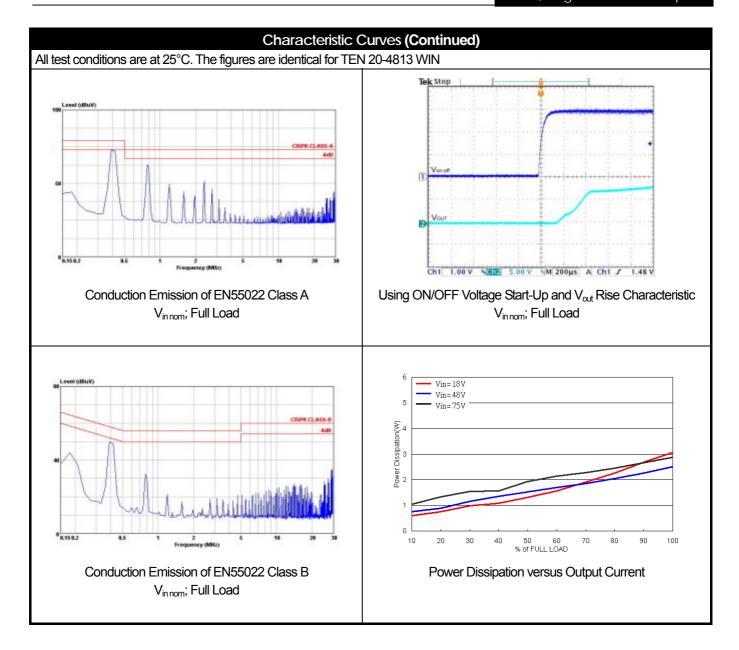


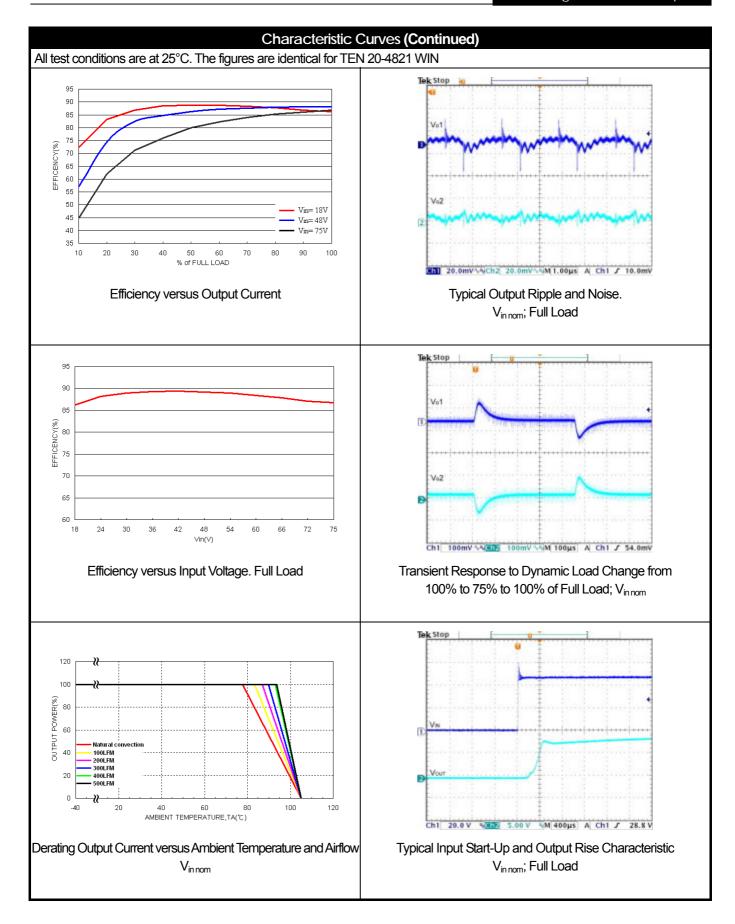
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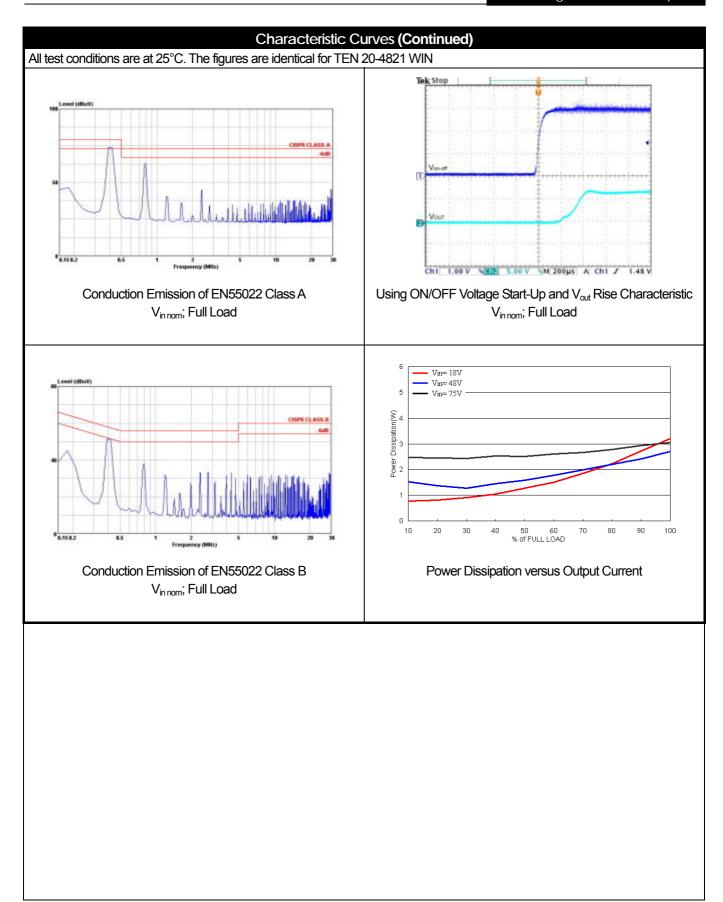


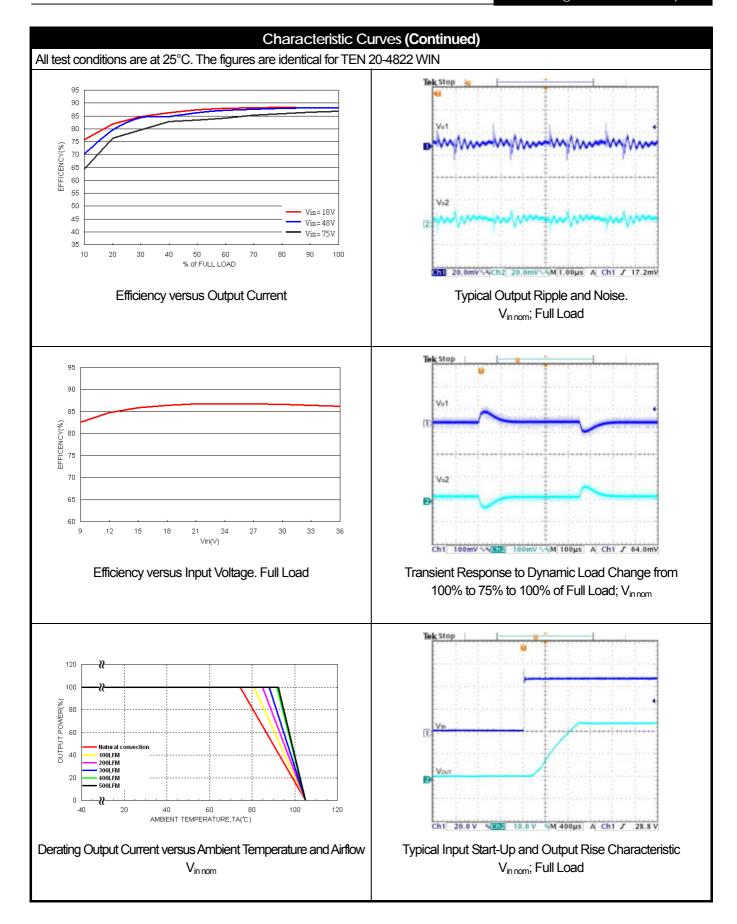


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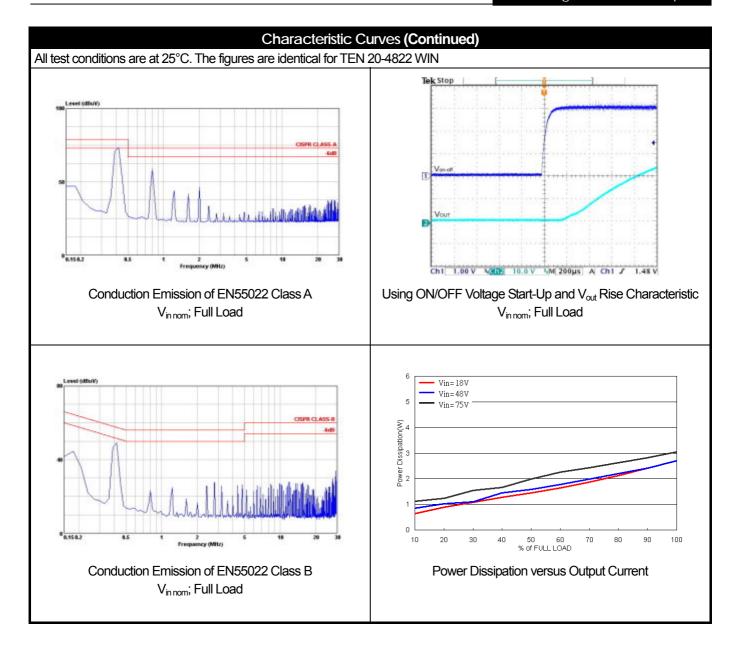


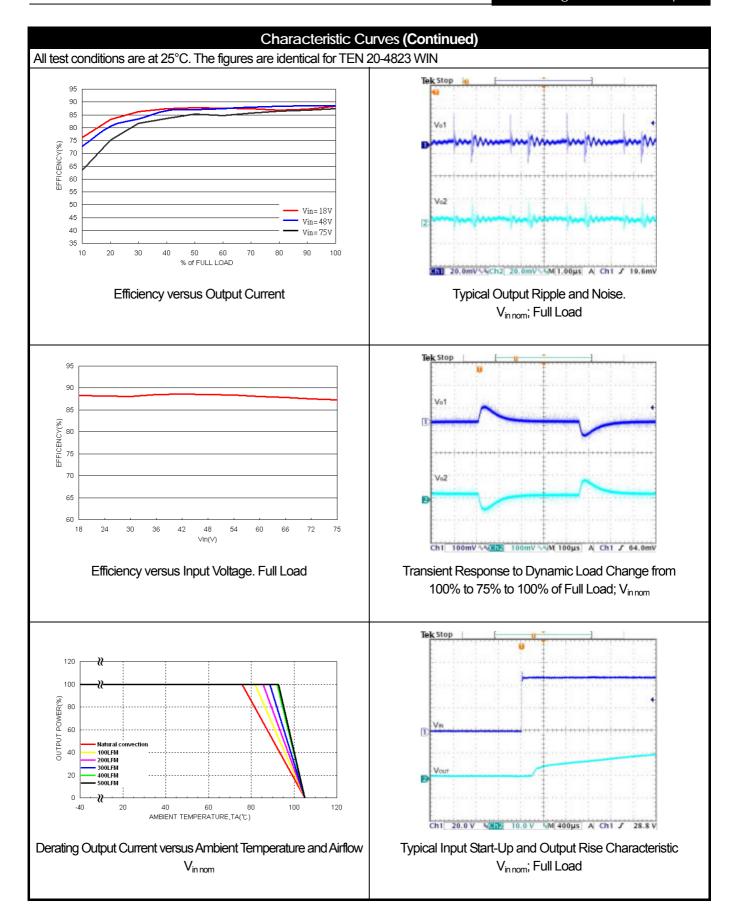




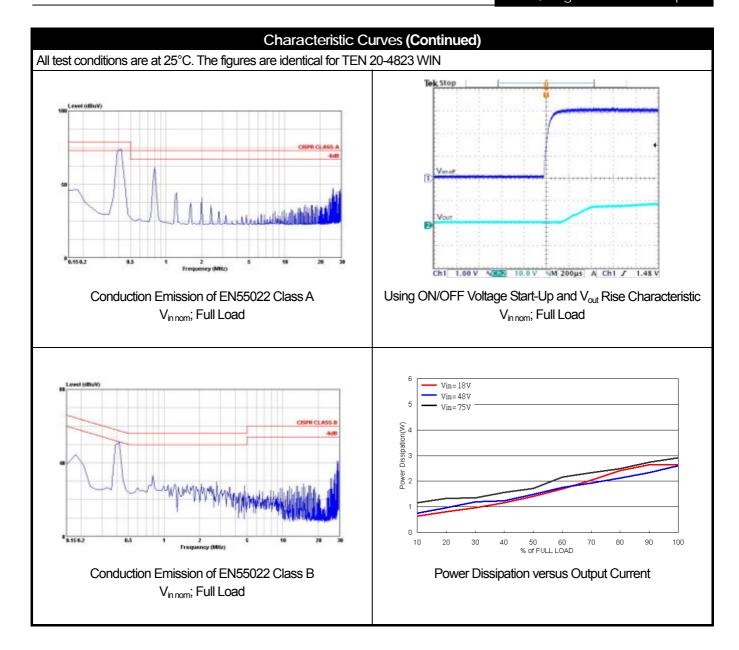


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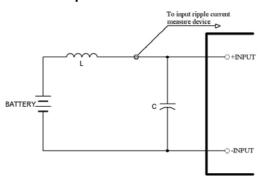


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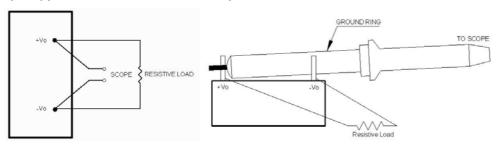
Testing Configurations

Input reflected-ripple current measurement test up

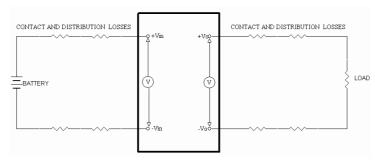


Component	Value	Voltage	Reference
L	12µH		
С	100µF	100V	Aluminum Electrolytic Capacitor

Peak-to-peak output ripple & noise measurement test up



Output voltage and efficiency measurement test up



Note: All measurements are taken at the module terminals.

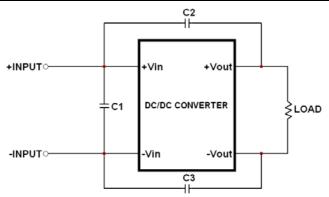
Single Output

$$Efficiency = \left(\frac{V_{out} \times I_{out}}{V_{in} \times I_{in}}\right) \times 100\%$$

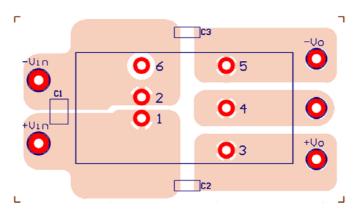
Dual Output

$$Efficiency = \left(\frac{V_{out1} \times I_{out1} + V_{out2} \times I_{out2}}{V_{\cdot} \times I_{\cdot}}\right) \times 100\%$$

EMC considerations Single Output



Suggested schematic for EN55022 conducted emission Class A



Recommended layout with input filter

To comply with conducted emissions noise EN55022 CLASS A following components are needed:

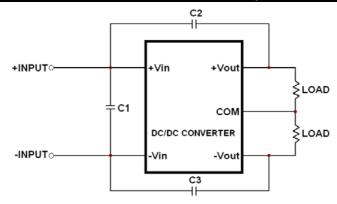
TEN 20-241x WIN

Component	Value	Voltage	Reference
C1			
C2, C3	1000pF	2KV	1808 MLCC

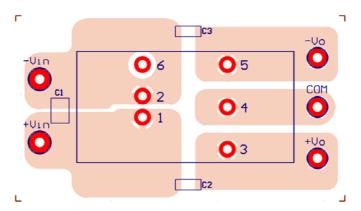
TEN 20-481x WIN

Component	Value	Voltage	Reference
C1	1µF	100V	1812 MLCC
C2, C3	1000pF	2KV	1808 MLCC

EMC considerations Dual Output



Suggested schematic for EN55022 conducted emission Class A



Recommended layout with input filter

To comply with conducted emissions noise EN55022 CLASS A following components are needed:

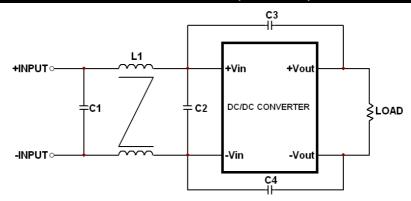
TEN 20-242x WIN

Component	Value	Voltage	Reference
C1			
C2, C3	1000pF	2KV	1808 MLCC

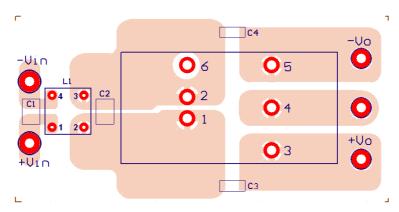
TEN 20-482x WIN

Component	Value	Voltage	Reference
C1	1µF	100V	1812 MLCC
C2, C3	1000pF	2KV	1808 MLCC

EMC considerations (Continued)



Suggested schematic for EN55022 conducted emission Class B



Recommended layout with input filter

To meet conducted emissions (EN55022 CLASS B) following components are needed:

TEN 20-241x WIN

Component	Value	Voltage	Reference
C1	4.7µF	50V	1812 MLCC
C3, C4	1000pF	2KV	1808 MLCC
L1	450µH		Common Choke, P/N: TCK-048

TEN 20-481x WIN

Component	Value	Voltage	Reference
C1, C2	2.2µF	100V	1812 MLCC
C3, C4	1000pF	2KV	1808 MLCC
L1	325µH		Common Choke, P/N: TCK-050

This Common Choke L1 has been define as follows:

■ TCK-048

L: $450\mu H \pm 35\%$ / DCR: $25m\Omega$, max

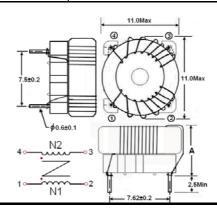
A height: 9.8 mm, Max

■ TCK-050

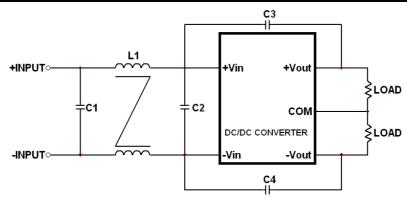
L: $325\mu H \pm 35\%$ / DCR: $35m\Omega$, max

A height: 8.8 mm, Max

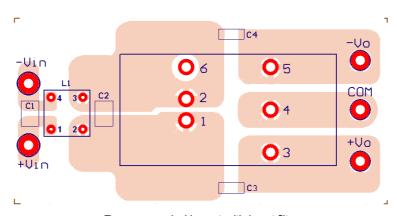
- Test condition: 100KHz / 100mV
- Recommended through hole: Ф0.8mm
- All dimensions in millimeters



EMC considerations (Continued)



Suggested schematic for EN55022 conducted emission Class B limits



Recommended layout with input filter

To comply with conducted emissions noise (EN55022 CLASS B) following components are needed:

TEN 20-242x WIN

Component	Value	Voltage	Reference
C1	4.7µF	50V	1812 MLCC
C3, C4	1000pF	2KV	1808 MLCC
L1	450µH		Common Choke, P/N: TCK-048

TEN 20-482x WIN

Component	Value	Voltage	Reference
C1, C2	2.2µF	100V	1812 MLCC
C3, C4	1000pF	2KV	1808 MLCC
L1	325µH		Common Choke, P/N: TCK-050

This Common Choke L1 has been define as follows:

■ TCK-048

L: $450\mu H \pm 35\%$ / DCR: $25m\Omega$, max

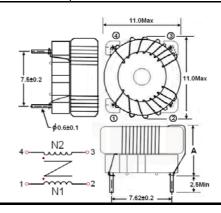
A height: 9.8 mm, Max

■ TCK-050

L: $325\mu H \pm 35\%$ / DCR: $35m\Omega$, max

A height: 8.8 mm, Max

- Test condition: 100KHz / 100mV
- Recommended through hole: Ф0.8mm
- All dimensions in millimeters



used to prevent those power devices from being damaged.

Input Source Impedance

The power module should be connected to a low impedance input source. Highly inductive source impedance can affect the stability of the power module. Input external L-C filter is recommended to minimize input reflected ripple current. The inductor has a simulated source impedance of $12\mu H$ and capacitor is a $220\mu F/100V$ low ESR type. The capacitor must be equipped as close as possible to the input terminals of the power module for lower impedance.

Output Over Current Protection

When excessive output currents occur in the system, circuit protection is required on all power supplies. Normally, overload current is maintained at approximately 140 percent of rated current for TEN 20-WIN series.

Hiccup-mode is a method of operation in a power supply whose purpose is to protect the power supply from being damaged during an over-current fault condition. It also enables the power supply to restart when the fault is removed. There are other ways of protecting the power supply when it is over-loaded, such as the maximum current limiting or current foldback methods. One of the problems resulting from over current is that excessive heat may be generated in power devices; especially MOSFET and Shottky diodes and the temperature of those devices may exceed their specified limits. A protection mechanism has to be

The operation of hiccup is as follows. When the current sense circuit sees an over-current event, the controller shuts off the power supply for a given time and then tries to start up the power supply again. If the over-load condition has been removed, the power supply will start up and operate normally; otherwise, the controller will see another over-current event and shut off the power supply again, repeating the previous cycle. Hiccup operation has none of the drawbacks of the other two protection methods, although its circuit is more complicated because it requires a timing circuit. The excess heat due to overload lasts for only a short duration in the hiccup cycle, hence the junction temperature of the power devices is much lower.

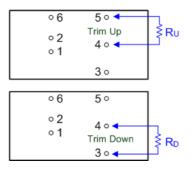
The hiccup operation can be done in various ways. For example, one can start hiccup operation any time an over-current event is detected; or prohibit hiccup during a designated start-up is usually larger than during normal operation and it is easier for an over-current event is detected; or prohibit hiccup during a designated start-up interval (usually a few milliseconds). The reason for the latter operation is that during start-up, the power supply needs to provide extra current to charge up the output capacitor. Thus the current demand during start-up is usually larger than during normal operation and it is easier for an over-current event to occur. If the power supply starts to hiccup once there is an over-current, it might never start up successfully. Hiccup mode protection will give the best protection for a power supply against over current situations, since it will limit the average current to the load at a low level, so reducing power dissipation and case temperature in the power devices.

Output Over Voltage Protection

The output over-voltage protection consists of output Zener diode that monitors the voltage on the output terminals. If the voltage on the output terminals exceeds the over-voltage protection threshold, then the Zener diode clamps the output voltage.

Output Voltage Adjustment

Output voltage set point adjustment allows the user to increase or decrease the output voltage set point of a module. This is accomplished by connecting an external resistor between the TRIM pin and either the $+V_{out}$ pin or $-V_{out}$ pin. With an external resistor between the TRIM and $-V_{out}$ pin, the output voltage set point increases. With an external resistor between the TRIM and $+V_{out}$ pin, the output voltage set point decreases.



TRIM TABLE

TEN 20-xx10 WIN

Trim up (%)	1	2	3	4	5	6	7	8	9	10
V _{OUT} (Volts) =	3.333	3.366	3.399	3.432	3.465	3.498	3.531	3.564	3.597	3.630
R _U (KΩ) =	57.930	26.165	15.577	10.283	7.106	4.988	3.476	2.341	1.459	0.753
Trim down (%)	1	2	3	4	5	6	7	8	9	10
V _{OUT} (Volts) =	3.267	3.234	3.201	3.168	3.135	3.102	3.069	3.036	3.003	2.970
$R_D(K\Omega) =$	69.470	31.235	18.490	12.117	8.294	5.745	3.924	2.559	1.497	0.647

TEN 20-xx11 WIN

Trim up (%)	1	2	3	4	5	6	7	8	9	10
V _{OUT} (Volts) =	5.050	5.100	5.150	5.200	5.250	5.300	5.350	5.400	5.450	5.500
R _U (KΩ) =	36.570	16.580	9.917	6.585	4.586	3.253	2.302	1.588	1.032	0.588
Trim down (%)	1	2	3	4	5	6	7	8	9	10
V _{OUT} (Volts) =	4.950	4.900	4.850	4.800	4.750	4.700	4.650	4.600	4.550	4.500

TEN 20-xx12 WIN

Trim up (%)	1	2	3	4	5	6	7	8	9	10
V _{OUT} (Volts)	= 12.120	12.240	12.360	12.480	12.600	12.720	12.840	12.960	13.080	13.200
$R_U(K\Omega)$	= 367.908	165.954	98.636	64.977	44.782	31.318	21.701	14.488	8.879	4.391
Trim down (%) 1	2	3	4	5	6	7	8	9	10
Trim down (%	4	2 11.760	Ť	4 11.520	5 11.400	6 11.280	7 11.160	8 11.040	9 10.920	10 10.800

TEN 20-xx13 WIN

Trim up (%)	1	2	3	4	5	6	7	8	9	10
V _{OUT} (Volts) =	15.150	15.300	15.450	15.600	15.750	15.900	16.050	16.200	16.350	16.500
$R_{U}(K\Omega) =$	404.184	180.592	106.061	68.796	46.437	31.531	20.883	12.898	6.687	1.718
Trim down (%)	4	2	2		_	_	_		•	40
THIT GOWII (78)		2	3	4	5	6	7	8	9	10
V_{OUT} (Volts) =			!	· -	14.250	14.100	13.950	13.800	13.650	13.500

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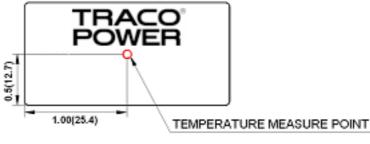
Short Circuitry Protection

Continuous, hiccup and auto-recovery mode.

During short circuit, converter still shut down. The average current during this condition will be very low and the device can be safety in this condition.

Thermal Consideration

The power module operates in a variety of thermal environments. However, sufficient cooling should be provided to help ensure reliable operation of the unit. Heat is removed by conduction, convection, and radiation to the surrounding Environment. Proper cooling can be verified by measuring the point as the figure below. The temperature at this location should not exceed 105°C. When Operating, adequate cooling must be provided to maintain the test point temperature at or below 105°C. Although the maximum point Temperature of the power modules is 105°C, you can limit this Temperature to a lower value for extremely high reliability.



Measurement shown in inches (mm)

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Remote ON/OFF Control

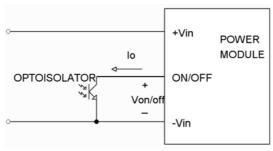
The positive logic remote On/Off control circuit is included.

Turns the module On during a logic High on the On/Off pin and turns Off during a logic Low.

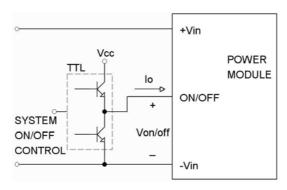
The On/Off pin is an open collector/drain logic input signal (Von/off) that referenced to GND.

If not using the remote on/off feature, please open circuit between on/off pin and -V_{in} pin to turn the module on.

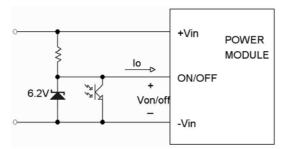
Remote On/Off Implementation



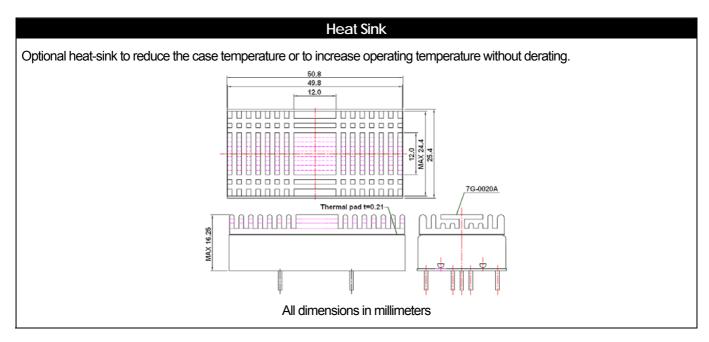
Isolated-Closure Remote On/Off

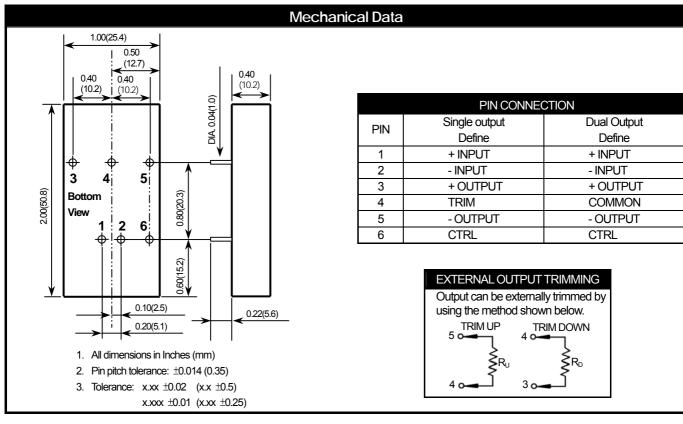


Level Control Using TTL Output

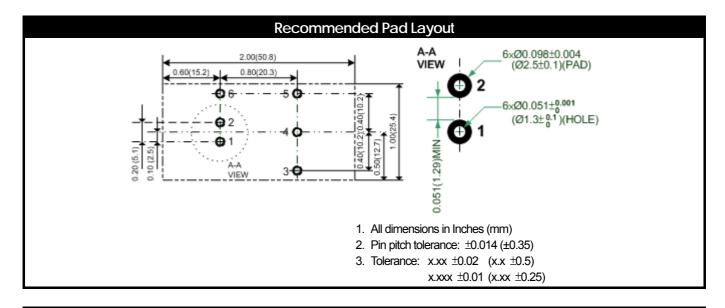


Level Control Using Line Voltage





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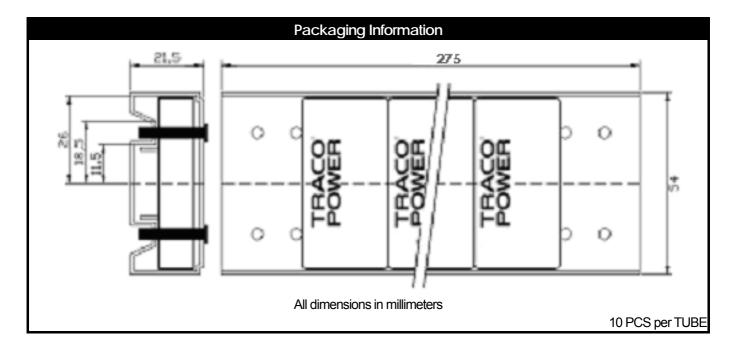


Soldering and Reflow Considerations

Reference Solder: Sn-Ag-Cu; Sn-Cu

Hand Welding: Soldering iron: Power 90W

Welding Time: 2~4 sec Temperature: 380~400°C



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Safety and Installation Instruction

Fusing Consideration

Caution: This power module is not internally fused. An input line fuse must always be used.

This encapsulated power module can be used in a wide variety of applications, ranging from simple stand-alone operation to an integrated part of sophisticated power architecture. To maximum flexibility, internal fusing is not included; however, to achieve maximum safety and system protection, always use an input line fuse. The safety agencies require a normal-blow fuse with maximum rating of 6A. Based on the information provided in this data sheet on Inrush energy and maximum dc input current; the same type of fuse with lower rating can be used. Refer to the fuse manufacturer's data for further information.

MTBF and Reliability

The MTBF of TEN 20-WIN series of DC/DC converters has been calculated according to:

Bellcore TR-NWT-000332 Case I: 50% stress, Operating Temperature at 40°C (Ground fixed and controlled environment). The resulting figure for MTBF is 1'691'000 hours.

MIL-HDBK 217F NOTICE2 FULL LOAD, Operating Temperature at 25°C. The resulting figure for MTBF is 629'000 hours.